Synthetic Tracking Controller for Robot Manipulator with Flexible Joints, Dynamics of Executive Motors and Affect of Disturbance Based on Radial Basic Function (RBF) Neural Network

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Abstract— This paper presents synthetic checking controller for robot manipulator with flexible joints, dynamics of executive motors and affect of disturbance based on radial basic function (RBF) neural network. A control algorithm is synthesized basing on the combination of backsteppping and RBF neural network in order to approximate the unknown functions. Finally, simulation results of a manipulator robot with single flexible joints based on Matlab-Simulink are presented to demonstrate the effectiveness of the proposed control algorithms

I. INTRODUCTION

S ynthesizing controller for robot manipulator with the assumption that links beten joints are flexible has received considerably research attention in recent years [1,2,3,4,5,6]. In [2], the problem of designing robust tracking controllers for a class of robotic manipulators with flexible joints, only using position measurements is developed with assumption that the input signal of the system is the torque generated by the executive motor. In this paper, control algorithms are synthesized to ensure the tracking error converged to zero after a certain time.

Several approaches have been introduced to solve the control problem of flexible-joint robots without including the actuator dynamics.

In [1], the researchers synthesize the control algorithms for robot manipulators with flexible joints based on sliding mode control. In fact, the simulated results have shod higher quality of control parameters with torque input generated from the executive motor.

As can be seen, torque placed on the joints is always generated by motors such as DC motors, stepper motors, BLDC motors and others. Hover, the assumption that the input signal of the manipulator robot is torque placed on the joints will create higher difficulty in practical control algorithms. Moreover, manipulators include uncertain parameters, therefore synthesizing controller with assumption of certain parameters is not accurate in real performance. Thus, the above research basically has achieved certain results in terms of control theory. Hover, realizing the above algorithms is huge obstacle in fact.

In order to apply these control algorithms, the synthesized controller must have signal inputs that voltage is placed on executive motors and the torque is placed on these joints. The performance of motor will depend on the voltage as ll as suitable methodology which need to value uncertain parameters. One of most effective solution is based on the adaptive control law.

Hover, when considering the dynamics of motor, the order of system will increase considerably. Therefore, not only the synthesis of control algorithm will be more complicated, but also the system with torque placed the joints can be not stable. Furthermore, when uncertain parameters are included on the manipulator, it will be quite difficult to synthesize the control algorithms

In recent years, the analytical study of adaptive nonlinear control systems using RBF universal function approximation has received much attention, typically, these methods are mentioned in documents [6–11].

The RBF network adaptation can effectively improve the control performance against large uncertainty of the system. The adaptation law is derived from the Lyapunov method so that the stability of the entire system and the convergence of the ight adaptation are guaranteed.

By using RBF control, significant improvement has been achieved when the system is subjected to a sudden change with system uncertainty.

Past research of universal approximation theorems on RBF [12,13] is shown that any nonlinear function with arbitrary accuracy can be approximated by RBF neural network over a compact set.

In order to realize these control algorithms in practice, this paper focuses on synthesizing control algorithm for robot manipulator, taking into account flexibility of joints and dynamics of executive motors. The methodology of Backstepping control combining with and based on basic radial function (BRF) neural network is also applied on this algorithm. The result is proven by simulation on Matlab – Simulink environment.

This paper consists five sections. Section I demonstrates the state equations and setting tracking issues. Section II represents the backstepping controller design. Section III focuses on synthesizing adaptive Backstepping control with RBF for single-link flexible joint robot. The estimated simulation is made in Section IV. In Section V, the conclusions are given.

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II. MODEL DESCRIPTION AND PROBLEM STATEMENT

2.1. Establishing state model and tracking problem

The mathematical model of a flexible-joint manipulator with DC excited permanent magnet motor can be described by

$$J_{l}(q_{1})\ddot{q}_{1} + C(q_{1},\dot{q}_{1})\dot{q}_{1} + G(q_{1}) = k_{e}(q_{2} - q_{1}) + d_{1}$$

$$J_{r}(q_{2})\ddot{q}_{2} + \mu(q_{2},\dot{q}_{2})\dot{q}_{2} + k_{e}(q_{2} - q_{1}) = K_{i}I_{u} + d_{2}$$

$$L_{u}\dot{I}_{u} + R_{u}I_{u} + K_{b}\dot{q}_{2} = u$$
(1)

From the mathematical model, the dynamic structure of flexible-joint robot manipulator with dc excited permanent magnet motors is shown in figure 1

For the convenience of representations, the state variables are used as below:

$$x_{1} = q_{1}; x_{2} = \dot{q}_{1}; x_{3} = q_{2}; x_{4} = \dot{q}_{2}; x_{5} = I_{u}$$
(2)
Where $x_{m} = [x_{m1}, x_{m2}...x_{mn}]^{T}, m = 1, 2..5.$

Substituting Equation (2) into Equation (1) yields and through some simple transformations obtain;

2.2. Backstepping controller design

Regarding to the main aim of adaptive backstepping sliding controller design for single-link flexible joint robot manipulator proposed in [4], controller is designed in several steps as follows:

Step 1: $e_1 = x_1 - x_{1d}$ is error beten output value and desired trajectories in tracking problem

Differential expression of above equation is as below

$$\dot{e}_{1} = \dot{x}_{1} - \dot{x}_{1d} = x_{2} - \dot{x}_{1d} \tag{4}$$

To realize $e_1 \rightarrow 0$, define a Lyapunov function as

$$V_{1} = \frac{1}{2}e_{1}^{2}$$
(5)

Then
$$\dot{V}_1 = e_1 \dot{e}_1 = e_1 (x_2 - \dot{x}_{1d})$$
 (6)

To realize $\dot{V_1} < 0$, if choose $x_2 = -k_1 e_1 + \dot{x}_{1d}, k_1 > 0$,

then
$$V = -k_1 e_1^2$$

Step 2: To realize $x_2 = -k_1e_1 + \dot{x}_{1d}$, using virtual control as $x_{2d} = -k_1 e_1 + \dot{x}_{1d}$ (7)

To realize
$$x_2 \rightarrow x_{2d}$$
, getting a new error $e_2 = x_2 - x_{2d}$

So
$$\dot{e}_2 = \dot{x}_2 - \dot{x}_{2d}$$
 (8)
Then $\dot{e}_2 = x_2 + g(x) - \dot{x}_2$ (9)

Then
$$\dot{e}_2 = x_3 + g(x) - \dot{x}_{2d}$$
 (9)

Then
$$\dot{V_1} = e_1(x_{2d} + e_2 - \dot{x}_{1d}) = -k_1 e_1^2 + e_2 e_1$$
 (10)



Fig. 1. Dynamic structure diagram of flexible joint robot manipulator with dc excited permanent magnets motor

,

$$\begin{aligned} \dot{x}_{1} &= x_{2} \\ \dot{x}_{2} &= x_{3} + g(x) \\ \dot{x}_{3} &= x_{4} \end{aligned} \qquad (3) \end{aligned} \qquad To realize $e_{2} \rightarrow 0 \text{ and } e_{1} \rightarrow 0 \text{, using a Lyapunov function as:} \\ \dot{x}_{2} &= x_{3} + g(x) \\ \dot{x}_{3} &= x_{4} \end{aligned} \qquad (3) \end{aligned} \qquad \begin{aligned} V_{2} &= V_{1} + \frac{1}{2}e_{2}^{2} = \frac{1}{2}(e_{1}^{2} + e_{2}^{2}) \\ \dot{x}_{4} &= x_{5} + h(x) \\ \dot{x}_{5} &= f(x) + bu \\ \text{Where} \end{aligned} \qquad \qquad (11) \end{aligned} \qquad To realize $\dot{V}_{2} &= -k_{1}e_{1}^{2} + e_{2}e_{1} + e_{2}(x_{3} + g(x) - \dot{x}_{2d}) \\ (11) \end{aligned} \qquad To realize \dot{V}_{2} &< 0, \text{ selecting} \\ x_{3} + g(x) - \dot{x}_{2d} &= -e_{1} - k_{2}e_{2} \\ (12) \end{aligned} \qquad Then \dot{V}_{2} &= -k_{1}e_{1}^{2} - k_{2}e_{2}^{2} \end{aligned} \qquad (13) \end{aligned}$$$$

$$f(x) = -\frac{R_u}{L_u} x_5 - \frac{K_B}{L_u} x_4; \ b = \frac{1}{L_u}$$

The purpose is to find the structure of in

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The purpose is to find the structure of input signal to meet the expected requirement that the output position converges to the desired reference trajectories $(x_1 \rightarrow x_{1d})$, with assumptions of existence of high order derivative of reference trajectories

Obviously, the output of the previous block is input of the next block in the cascade system [3]. With this cascade property, backstepping synthesis method can be applied easily.

$$V_2 = V_1 + \frac{1}{2}e_2^2 = \frac{1}{2}(e_1^2 + e_2^2)$$

Then
$$\dot{V}_2 = -k_1 e_1^2 + e_2 e_1 + e_2 (x_3 + g(x) - \dot{x}_{2d})$$
 (11)

To realize $V_2 < 0$, selecting

$$\dot{x}_3 + g(x) - \dot{x}_{2d} = -e_1 - k_2 e_2$$
 (12)

Then
$$\dot{V}_2 = -k_1 e_1^2 - k_2 e_2^2$$
 (13)

tep 3: To realize $x_3 + g(x) - \dot{x}_{2d} = -e_1 - k_2 e_2$, selecting virtual control as

$$x_{3d} = -e_1 - k_2 e_2 - g(x) + \dot{x}_{2d}$$
(14)
To realize $x_3 \to x_{3d}$, defining the error

$$10 \text{ feature} \quad x_3 \quad x_{3d}, \text{ defining the effor}$$

$$e_3 - x_3 - x_{3d} \tag{15}$$

(15)

$$\begin{array}{l} \text{nave } e_3 = x_3 - x_{3d} = x_4 - x_{3d} \end{array} \tag{10}$$

To realize $e_3 \rightarrow 0$ and $e_2 \rightarrow 0$ and $e_1 \rightarrow 0$, choose a Lyapunov function as

$$V_3 = V_2 + \frac{1}{2}e_3^2 = \frac{1}{2}(e_1^2 + e_2^2 + e_3^2)$$
(17)

Substituting Equations (11) and (16) into (17) yields

$$\dot{V}_3 = -k_1 e_1^2 - k_2 e_2^2 + e_3 (e_2 + x_4 - \dot{x}_{3d})$$
 (18)
To realize $\dot{V}_3 < 0$, choose

 $e_2 + x_4 - \dot{x}_{3d} = -k_3 e_3, \, k_3 > 0 \tag{19}$

Then $\dot{V}_3 = -k_1e_1^2 - k_2e_2^2 - k_3e_3^2$

Step 4: To realize (19), choose virtual control as $x = \dot{x} - ke - e$ (20)

$$x_{4d} - \dot{x}_{3d} - \dot{x}_{3}e_{3} - e_{2}$$
To realize $x_{4} \to x_{4d}$, define a new error $e_{4} = x_{4} - x_{4d}$
have $\dot{e}_{4} = \dot{x}_{4} - \dot{x}_{4d} = x_{5} + h(x) - \dot{x}_{4d}$
(21)

To realize $e_4 \to 0$ and $e_3 \to 0$ and $e_2 \to 0$ and $e_1 \to 0$, define a Lyapunov function as

$$V_4 = V_3 + \frac{1}{2}e_4^2 = \frac{1}{2}(e_1^2 + e_2^2 + e_3^2 + e_4^2)$$

Then
 $\dot{V}_4 = b_1e_2^2 + b_2e_3^2 + b_3e_4^2 + b_4e_4^2$

$$\dot{V}_4 = -k_1 e_1^2 - k_2 e_2^2 - k_3 e_3^2 + e_4 (e_3 + x_5 + h(x) - \dot{x}_{4d})$$
(22)
To realize $\dot{V}_4 < 0$, choose

$$e_3 + x_5 + h(x) - \dot{x}_{4d} = -k_4 e_4$$
Then $\dot{V}_4 = -k_1 e_1^2 - k_2 e_2^2 - k_3 e_3^2 - k_4 e_4^2$
(23)

Step 5: To realize (23), choose virtual control as

$$x_{5d} = \dot{x}_{4d} - k_4 e_4 - h(x) - e_3$$
To realize $x_5 \rightarrow x_{5d}$, get a new error $e_5 = x_5 - x_{5d}$
(24)

have
$$\dot{e}_{5} = \dot{x}_{5} - \dot{x}_{5d} = f(x) + bu - \dot{x}_{5d}$$
 (25)

To realize $e_5 \to 0$ and $e_4 \to 0$ and $e_3 \to 0$ and $e_2 \to 0$ and $e_1 \to 0$, define a Lyapunov function as

$$V_{5} = V_{4} + \frac{1}{2}e_{5}^{2} = \frac{1}{2}(e_{1}^{2} + e_{2}^{2} + e_{3}^{2} + e_{4}^{2} + e_{5}^{2}), \text{ Then}$$

$$\dot{V}_{5} = -k_{1}e_{1}^{2} - k_{2}e_{2}^{2} - k_{3}e_{3}^{2} - k_{4}e_{4}^{2} + e_{5}(e_{4} + f(x) + bu - \dot{x}_{5d})$$

To realize $\dot{V}_{5} < 0$, choose

$$e_{4} + f(x) + bu - \dot{x}_{5d} = -k_{5}e_{5}, k_{5} > 0$$
Then $\dot{V}_{5} = -k_{1}e_{1}^{2} - k_{2}e_{2}^{2} - k_{3}e_{3}^{2} - k_{4}e_{4}^{2} - k_{5}e_{5}^{2}$

$$(26)$$

From (26) can get the control law:

$$u = \frac{1}{b}(-f(x) + \dot{x}_{5d} - k_5 e_5 - e_4)$$
(27)

To realize (27), need to obtain information about g(x), h(x), f(x). For manipulators, the above functions consist of uncertain parameters, therefore in the practice it is hard to get above information from objective model. In order to have the information on those functions, a method approximated them is used necessarily. 2.3. Adaptive backstepping control with RBF for single link flexible joint robot manipulation



Fig. 2. Block diagram of the adaptive backstepping -RBF control

In (3), the functions g(x), h(x), f(x) and b are unknown, the bound of b is known $b \ge \zeta(\zeta > 0)$. The unknown functions of $\hat{g}, \hat{h}, \hat{f}$ can be approximated by neural network and \hat{b} can be estimated by adaptive law as in figure 2

2.3.1. Backstepping Controller Design estimating unknown Functions

Refer to the main idea of adaptive neural network backstepping sliding controller design for single-link flexible joint robot, several steps of designing controller are as follows:

Step 1: Similar as above,

$$\dot{V}_1 = e_1 \dot{e}_1 = e_1 (x_2 - \dot{x}_{1d})$$
 (28)
To realize $\dot{V}_1 < 0$, if choose $x_2 = -k_1 e_1 + \dot{x}_{1d}, k_1 > 0$

Step 2:

$$\dot{V}_2 = -k_1 e_1^2 + e_2 e_1 + e_2 (x_3 + g(x) - \dot{x}_{2d})$$
 (29)

To realize $\dot{V} < 0$,

$$x_3 + g(x) - \dot{x}_{2d} = -e_1 - k_2 e_2$$
(30)
Step 3 Similar as above,

$$\dot{V}_{3} = -k_{1}e_{1}^{2} - k_{2}e_{2}^{2} + e_{2}(g(x) - \hat{g}(x)) + e_{3}(e_{2} + x_{4} - \dot{x}_{3d})$$
(31)

To realize $\dot{V}_3 < 0$,

$$e_2 + x_4 - \dot{x}_{3d} = -k_3 e_3, \, k_3 > 0 \tag{32}$$

Then $\dot{V}_3 = -k_1 e_1^2 - k_2 e_2^2 - k_3 e_3^2 + e_2(g(x) - \hat{g}(x))$ **Step 4:** To realize $e_2 + x_4 - \dot{x}_{3d} = -k_3 e_3$, choose virtual control as $x_{4d} = \dot{x}_{3d} - k_3 e_3 - e_2$ (33) Similar as above

$$\dot{V}_{4} = -k_{1}e_{1}^{2} - k_{2}e_{2}^{2} - k_{3}e_{3}^{2} + e_{2}(g(x) - \hat{g}(x)) + e_{4}(e_{3} + x_{5} + h(x) - \dot{x}_{4d})$$
(34)

To realize $\dot{V}_{4} < 0$,

$$e_3 + x_5 + h(x) - \dot{x}_{4d} = -k_4 e_4 \tag{35}$$

Then
$$\dot{V}_4 = -k_1 e_1^2 - k_2 e_2^2 - k_3 e_3^2 - k_4 e_4^2 + e_2 (g(x) - \hat{g}(x))$$

Step 5: To realize $e_3 + x_5 + h(x) - \dot{x}_{4d} = -k_4 e_4$

choose virtual control as

$$x_{5d} = \hat{\dot{x}}_{4d} - k_4 e_4 - \hat{h}(x) - e_3 \tag{36}$$

Where $\dot{x}_{4d} = \dot{x}_{4d1} - d$, $\hat{x}_{4d} = \dot{x}_{4d1} - \hat{d}$

 \dot{x}_{4d1} composes of known values, d consists of the unknown

parts, where d is estimated value of d, with

$$\dot{x}_{4d1} = -(k_2 + k_3)x_4 + (k_2k_3 - 2)x_3 - k_3x_2 + k_3\dot{x}_{1d} + \ddot{x}_{1d} + (k_2k_3 + 1)\dot{x}_{2d} + (k_2 + k_3 + 1)\ddot{x}_{2d}$$
(37)

and

$$d = (k_3k_2 + 2)g(x) + k_2\dot{g}(x) + k_3\dot{g}(x) + \ddot{g}(x)$$
(38)
To realize $x \rightarrow x$, define a new error $e = x - x$

To realize $x_5 \rightarrow x_{5d}$, define a new error $e_5 = x_5 - x_{5d}$ then

$$\dot{e}_{5} = \dot{x}_{5} - \dot{x}_{5d} = f(x) + bu - \dot{x}_{5d}$$
 (39)
where

$$\dot{x}_{5d} = \hat{x}_{4d} - k_4 e_4 - \hat{h}(x) - e_3 = \dot{x}_{5d1} + \dot{x}_{5d2}$$
(40)

 \dot{x}_{5d1} composes of known values, \dot{x}_{5d2} consists of the unknown parts. With

$$\dot{x}_{5d1} = -k_4 x_5 - (k_2 + k_3 - 1)x_4 - (k_3 + 1)x_2 + + (k_2 k_3 - k_2 - 2)x_3 + (k_3 + 1)\dot{x}_{1d} + \ddot{x}_{1d} + + (k_2 k_3 + k_2 + 1)\dot{x}_{2d} + (k_2 + k_3 + 2)\ddot{x}_{2d} + k_4 \dot{x}_{4d} \dot{x}_{5d2} = -k_4 h(x) - k_2 g(x) - \dot{g}(x) - \dot{d} - \dot{h}(x)$$
(41)

Define
$$\bar{f} = f - \dot{x}_{5d2}$$
.
 $\dot{e}_5 = \bar{f} - \dot{x}_{5d1} + (b - \hat{b})u + \hat{b}u$ (42)

Where \hat{b} is estimation value of b.

To realize $e_5 \to 0$ and $e_4 \to 0$ and $e_3 \to 0$ and $e_2 \to 0$ and $e_1 \to 0$. Similar as above,

$$\dot{V}_{5} = -k_{1}e_{1}^{2} - k_{2}e_{2}^{2} - k_{3}e_{3}^{2} - k_{4}e_{4}^{2} + e_{2}(g(x) - \hat{g}(x)) + + e_{4}(d - \hat{d}) + e_{4}(h(x) - \hat{h}(x)) + e_{5}(b - \hat{b})u + + e_{5}(e_{4} + \bar{f} - \dot{x}_{5d1} + \hat{b}u)$$
(43)

To realize $\dot{V}_5 < 0$, choose virtual control as

$$u = \frac{1}{\hat{b}} (-e_4 - \hat{\bar{f}} + \dot{x}_{5d1} - k_5 e_5)$$
(44)

Then

$$\dot{V}_{5} = -k_{1}e_{1}^{2} - k_{2}e_{2}^{2} - k_{3}e_{3}^{2} - k_{4}e_{4}^{2} - k_{5}e_{5}^{2} + e_{2}(g - \hat{g}) + e_{4}\left(d - \hat{d}\right) + e_{4}\left(h - \hat{h}\right) + e_{5}(\bar{f} - \hat{\bar{f}}) + e_{5}(b - \hat{b})u$$
(45)

If $b = \hat{b}, g = \hat{g}, h = \hat{h}, f = \hat{f}, d = \hat{d}$, get $\dot{V}_{5} < 0$ 2.3.2.. Backstepping Controller Design with RBF Approximation

To approximate the unknown function g(x), h(x), f(x), use RBF neural network as in below [5]:

$$\begin{cases} g = W_1^T h_1 + \varepsilon_1 \\ h = W_2^T h_2 + \varepsilon_2 \\ d = W_3^T h_3 + \varepsilon_3 \\ \overline{f} = W_4^T h_4 + \varepsilon_4 \end{cases}$$
(46)

Where W_i is the neural network weighted value, h_i is the Gaussian function,, ε_i is the approximation error and $\|\varepsilon\| = \left\| \left[\varepsilon_1 \ \varepsilon_2 \ \varepsilon_3 \ \varepsilon_4 \right]^T \right\| < \varepsilon_N$, $\|W\|_F \leq W_m$. Define

$$\begin{cases} \hat{g} = \hat{W}_1^T h_1 \\ \hat{h} = \hat{W}_2^T h_2 \\ d = \hat{W}_3^T h_3 \\ \overline{f} = \hat{W}_4^T h_4 \end{cases}$$
(47)

Where \hat{W}_i^T is weighted value estimation, Define

$$Z = \begin{bmatrix} 0 & & & \\ & W_1 & & \\ & & W_2 & & \\ & & & W_3 & \\ & & & & W_4 \end{bmatrix}, \|Z\|_F \le Z_M$$
(48)

$$\hat{Z} = \begin{vmatrix} 0 & & \\ \hat{W}_1 & & \\ & \hat{W}_2 & \\ & & \hat{W}_3 & \\ & & & \hat{W}_4 \end{vmatrix}, \quad \tilde{Z} = Z - \hat{Z}$$
(49)

Define a Lyapunov function as

$$V = \frac{1}{2}\xi^{T}\xi + \frac{1}{2}tr(\tilde{Z}^{T}\Gamma^{-1}\tilde{Z}) + \frac{1}{2}\eta\tilde{b}^{2}$$
(50)

Where $V_5 = \frac{1}{2}\xi^T \xi, \eta > 0, \Gamma$ is a positive-definite matrix with proper dimension

$$\Gamma = \begin{bmatrix} 0 & & & \\ & \Gamma_2 & & \\ & & \Gamma_3 & \\ & & & \Gamma_4 & \\ & & & & \Gamma_5 \end{bmatrix},$$
 and $\tilde{b} = b - \hat{b}$

Let adaptive law as

$$\dot{\hat{Z}} = \Gamma h \xi^{T} - n \Gamma \|\xi\| \hat{Z}$$
(51)

Where $h = \begin{bmatrix} 0 & h_1 & h_2 & h_3 & h_4 \end{bmatrix}^T$, *n* is a positive number, and $\hat{b}(0) \ge b$. Then have

$$\begin{split} \dot{V} &= \xi^T \dot{\xi} + tr(\tilde{Z}^T \Gamma^{-1} \dot{\tilde{Z}}) + \eta \tilde{b} \dot{\tilde{b}} = -k_1 e_1^2 - k_2 e_2^2 - k_3 e_3^2 - k_4 e_4^2 \\ &- k_5 e_5^2 + (\tilde{W}_1^T h_1 + \varepsilon_1) e_2 + (\tilde{W}_2^T h_2 + \varepsilon_2) e_4 + (\tilde{W}_3^T h_3 + \varepsilon_3) e_4 \\ &+ e_5 (\tilde{W}_4^T h_4 + \varepsilon_4) + tr(\tilde{Z}^T \Gamma^{-1} \dot{\tilde{Z}}) + \tilde{b} e_5 u + \eta \tilde{b} \dot{\tilde{b}} \end{split}$$

Where $\tilde{W}_{i}^{T} = W_{i}^{T} - \hat{W}_{i}^{T}$, i = 1, 2, 3, 4, then

$$\dot{V} = -\xi^{T} K_{e} \xi + \xi^{T} \varepsilon + tr(\tilde{Z}^{T} \Gamma^{-1} \tilde{Z} + \tilde{Z}^{T} h \xi^{T}) + \tilde{b} e_{5} u + \eta \tilde{b} \tilde{b}$$
Where $K_{e} = \begin{bmatrix} k_{1} & k_{2} & k_{3} & k_{4} & k_{5} \end{bmatrix}$ and $\varepsilon = \begin{bmatrix} 0 & \varepsilon_{1} & \varepsilon_{2} & \varepsilon_{3} & \varepsilon_{4} \end{bmatrix}^{T}$
Then $\dot{Z} = -\dot{Z}$, and $\dot{b} = \dot{b}$, from (51), have
 $\dot{V} = -\xi^{T} K_{e} \xi + \xi^{T} \varepsilon + n \|\xi\| tr(\tilde{Z}^{T} (Z - \tilde{Z})))$
 $+ \tilde{b} (e_{5} u - \eta \dot{b})$
(52)

To guarantee $\tilde{b}(e_5u + \eta \dot{\tilde{b}}) \le 0$, at the same time to avoid singularity in (44) and guarantee $\hat{b} > \underline{b}$, using an adaptive law for \hat{b} given in [6] is as follows:

 $\dot{\hat{b}} = \begin{cases} \eta^{-1} e_{5}u, \text{ if } e_{5}u > 0\\ \eta^{-1} e_{5}u, \text{ if } e_{5}u \le 0 \text{ and } \hat{b} > \underline{b}\\ \eta^{-1}, \text{ if } e_{5}u \le 0 \text{ and } \hat{b} \le \underline{b} \end{cases}$ (53)

The adaptive law (53) can be analyzed as: 1. If $e_{5}u > 0$, get $\tilde{b}(e_{5}u + \eta\dot{\tilde{b}}) = 0$ and $\dot{\tilde{b}} > 0$; thus $\hat{b} > \underline{b}$ 2. If $e_{5}u \le 0$ and $\hat{b} > \underline{b}$, get $\tilde{b}(e_{5}u + \eta\dot{\tilde{b}}) = 0$ 3. if $e_{5}u \le 0$ và $\hat{b} \le \underline{b}$ have $\tilde{b} = b - \hat{b} \ge b - \underline{b} > 0$; thus

 $\tilde{b}(e_{5}u - \eta \hat{b}) = \tilde{b}e_{5}u - \tilde{b} \leq 0$ and \hat{b} will increase gradually,

and then $\hat{b} > b$ s will be guaranteed with $\dot{\hat{b}} > 0$

According to Schwarz inequality, have

 $tr(\tilde{Z}^T(Z-\tilde{Z})) \le \|\tilde{Z}\|_F \|Z\|_F - \|\tilde{Z}\|_F^2$, then $K_{\min} \|\xi\|^2 \le \xi^T K \xi$, K_{\min} is the minimum eigenvalue of K, then (54) becomes

$$egin{aligned} \dot{V} &\leq -K_{\min} \left\| \xi
ight\|^2 + arepsilon_N \left\| \xi
ight\| + n \left\| \xi
ight\| \left(\left\| ilde{Z}
ight\|_F \left\| Z
ight\|_F - \left\| ilde{Z}
ight\|_F^2
ight) + M \ &\leq - \left\| \xi
ight\| \left(K_{\min} \left\| \xi
ight\| - arepsilon_N + n \left\| ilde{Z}
ight\|_F \left(\left\| ilde{Z}
ight\|_F
ight) - Z_M
ight) \end{aligned}$$

Since

$$\begin{split} K_{\min} \left\| \boldsymbol{\xi} \right\| &- \boldsymbol{\varepsilon}_{N} + n \Big(\left\| \tilde{\boldsymbol{Z}} \right\|_{F}^{2} - \left\| \tilde{\boldsymbol{Z}} \right\|_{F} \boldsymbol{Z}_{M} \Big) = K_{\min} \left\| \boldsymbol{\xi} \right\| - \\ &- \boldsymbol{\varepsilon}_{N} + n \Big(\left\| \tilde{\boldsymbol{Z}} \right\| - \frac{1}{2} \boldsymbol{Z}_{M} \Big)^{2} - \frac{n}{4} \boldsymbol{Z}_{M}^{2} \end{split}$$

this implies that $\dot{V} < 0$ as long as

$$\left\|\xi\right\| > \frac{\varepsilon_{\scriptscriptstyle N} + \frac{n}{4}Z_{\scriptscriptstyle M}^2}{K_{\scriptscriptstyle \min}} \text{ or } \left\|\tilde{Z}\right\|_{\scriptscriptstyle F} > \frac{1}{2}Z_{\scriptscriptstyle M} + \sqrt{\frac{Z_{\scriptscriptstyle M}^2}{4} + \frac{\varepsilon_{\scriptscriptstyle N}}{n}}$$

2.4. Simulation and Result remarks

2.4.1. Description of the Single-link robot manipulator with flexible joint

To evaluate the dynamic behaviour of the proposed control algorithm, simulation results for single-link flexiblejoint robots (figure 3) are presented to estimate the effectiveness of the proposed control algorithms.



Fig. 3. Single-link robot manipulator with flexible joint Considering the dynamics of the single-link manipulator with rotational joints as described by

$$J_{l}(q_{1})\ddot{q}_{1} + mgl\sin q_{1} = k_{e}(q_{2} - q_{1})$$

$$J_{r}\ddot{q}_{2} + \mu\dot{q}_{2} + k_{e}(q_{2} - q_{1}) = K_{i}\dot{i}_{u}$$

$$L_{u}\dot{I}_{u} + R_{u}I_{u} + K_{R}\dot{q}_{2} = u$$
(54)

TABLE I PHYSICAL PARAMETERS OF THE ROBOT MANIPULATOR



Fig. 4. The structure diagram of proposed controller The parameters of the proposed controller

$$k_1 = k_2 = k_3 = k_4 = k_5 = 3,5$$
; $\Gamma_2 = \Gamma_3 = \Gamma_4 = \Gamma_5 = 250$
 $c_i = [-1 - 0.5 \ 0 \ 0.5 \ 1], b_i = 1.5, \eta = 150$

The simulation results are shown in figs. 5-11, respectively





Fig. 11. Current value of DC motor

2.4.2. Result remarks

The simulation results for single-link flexible-joint robots based on proposed control algorithms are presented in figures 5-11. Several remarks can be seen as below:

- The angular position of joints was tracked to the desired reference trajectories with the error values converging towards 0 in a certain period of time.

- The values of unknown functions g(x), h(x), f(x) are approximated basically by RBF neural network, therefore this permit to build control algorithm with higher quality.

- From the achieved results, it is concluded that the proposed controller design brings to the expected results.

III. CONCLUSION

The study successfully establishes the synthesized controller for robot manipulator with flexible joints, based on radial basic function (RBF) neural network. This controller permit to control the joint positions of robot manipulator including dynamics of electric motors in order to achieving high-precision position control. Although the order of systems is higher, but the angular position of joints was tracked to the desired reference trajectories with the error values converge towards zero in a certain period of time.

From above results, this study can be extended to handle a broader class of complicated flexible-joint robots

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